TTM Technologies

RAD HARD ULTRA LOW DROPOUT ADJUSTABLE POSITIVE LINEAR REGULATOR 5910RH

FEATURES:

• Total Dose Hardened to 100 Krads(Si) (Method 1019.7 Condition A)
• Ultra Low Dropout for Reduced Power Consumption
• External Shutdown/Reset Function
• Latching Overload Protection
• Adjustable Output Using Two External Resistors
• Output Current Limit
• Surface Mount Package Available with Lead Forming
• Low Input Voltage for Maximum Efficiency
• Up to 5A Output Current
• Available as SMD 5962R05220
• RAD Certified by DLA
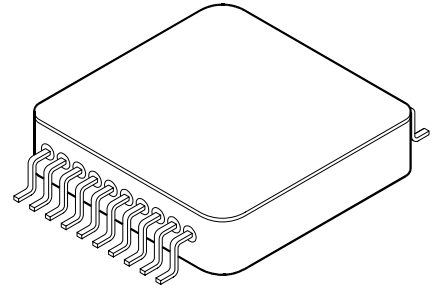
DESCRIPTION:

The MSK5910RH is a rad hard adjustable linear regulator capable of delivering 5.0 amps of output current. The typical dropout is only 0.11 volts at 1 amp. An external shutdown/reset function is ideal for power supply sequencing. This device also has latching overload protection that requires no external current sense resistor. The MSK5910RH is radiation hardened and specifically designed for many space/satellite applications. The device is packaged in a hermetically sealed 20 pin flatpack that can be lead formed for surface mount applications.

EQUIVALENT SCHEMATIC
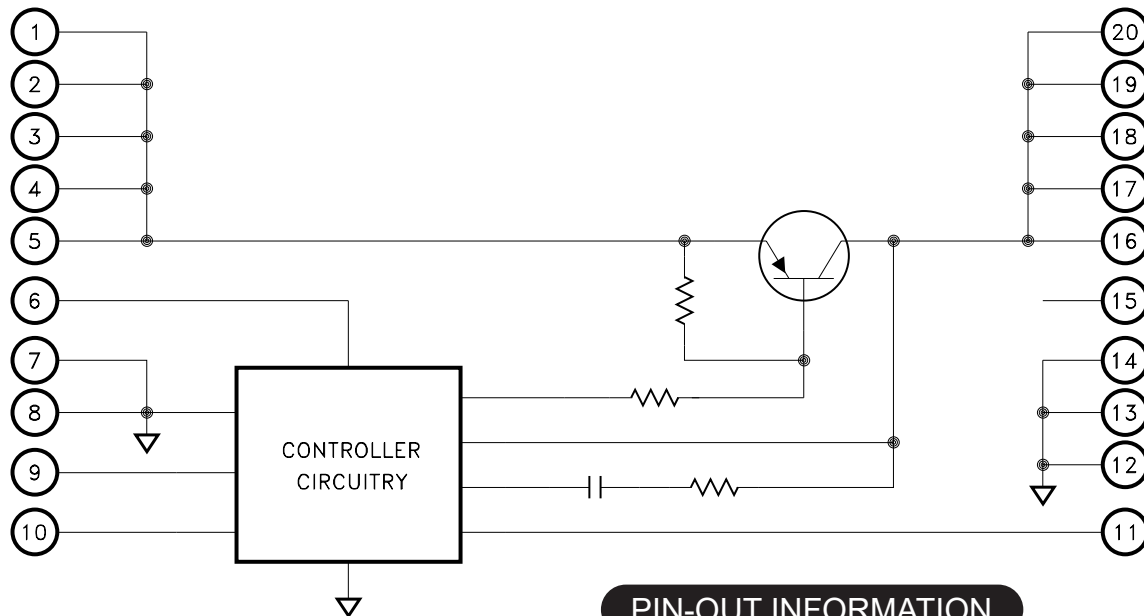
TYPICAL APPLICATIONS

• Satellite System Power Supplies
• Switching Power Supply Post Regulators
• Constant Voltage/Current Regulators
• Microprocessor Power Supplies

PIN-OUT INFORMATION

1	VINA	11	FB
2	VINB	12	GND2
3	VINC	13	GND2
4	VIND	14	GND2
5	VINE	15	NC
6	VBIAS	16	VOUTA
7	GND1	17	VOUTB
8	GND1	18	VOUTC
9	Latch	19	VOUTD
10	Shutdown	20	VOUTE

CASE = ISOLATED

ABSOLUTE MAXIMUM RATINGS ⑧

+V_{BIAS} Bias Supply Voltage 10V
+V_{IN}   Supply Voltage 10V
V_{SD}   Shutdown Voltage 10V
I_{OUT} Output Current ..... ⑦ 5A
T_C   Case Operating Temperature Range
        MSK5910K/H RH .... -55°C to +125°C
        MSK5910RH .... -40°C to +85°C

T_{ST} Storage Temperature Range .... -65°C to +150°C
T_{LD} Lead Temperature Range
       (10 Seconds) .... 300°C
P_D Power Dissipation .... See SOA Curve
T_J Junction Temperature .... 150°C

ELECTRICAL SPECIFICATIONS ⑩

Parameter	Test Conditions ① ⑨	Group A Subgroup	MSK5910K/H Min.	Typ.	Max.	MSK5910 Min.	Typ.	Max.	Units
Input Voltage Range ②	10mA ≤ I_{OUT} ≤ 1.0A	1,2,3	2.0	-	7.5	2.0	-	7.5	V
Input Bias Voltage ②	V_{BIAS} ≥ V_{IN}	1,2,3	2.9	5.0	7.5	2.9	5.0	7.5	V
Feedback Voltage	I_{OUT} = 1.0A R1 = 187Ω	1	1.225	1.265	1.305	1.202	1.265	1.328	V
		2,3	1.225	-	1.305	-	-	-	V
	Post Radiation	1	1.215	-	1.315	-	-	-	V
Feedback Pin Current ②	V_{FB} = 1.265V 10mA ≤ I_{OUT} ≤ 1.0A	1,2,3	0	-	5.0	0	-	5.0	µA
Quiescent Current	I_{IN} + I_{BIAS}, V_{BIAS} = V_{IN} = 7.5V Not Including I_{OUT}	1,2,3	-	14	20	-	14	20	mA
Bias Current	V_{BIAS} = 7.5V	1,2,3	-	2	4	-	2	4	mA
Line Regulation	I_{OUT} = 10mA 2.9V ≤ V_{IN} ≤ 7.5V R1 = 187Ω	1	-	±0.01	±0.50	-	0.01	±0.60	%VOUT
		2,3	-	-	±0.50	-	-	-	%VOUT
Load Regulation	10mA ≤ I_{OUT} ≤ 1.0A R1 = 976	1	-	±0.06	±0.80	-	0.06	±1.0	%VOUT
		2,3	-	-	±0.80	-	-	-	%VOUT
Dropout Voltage	Delta FB = 1% I_{OUT} = 1.0A	1	-	0.11	0.40	-	0.11	0.45	V
		2,3	-	0.14	0.40	-	-	-	V
Minimum Output Current ②	2.9V ≤ V_{IN} ≤ 7.5V R1 = 187Ω	1	-	8	10	-	8	10	mA
		2,3	-	9	10	-	-	-	mA
Output Voltage Range ②	V_{IN} = 7.5V	-	1.5	-	7.0	1.5	-	7.0	V
Output Current Limit ⑦	V_{IN} = 2.5V V_{OUT} = 1.5V	1	3.2	3.6	4.0	3.2	3.6	4.0	A
		2,3	3.0	-	-	-	-	-	A
Shutdown Threshold	V_{OUT} ≤ 0.2V (OFF) V_{OUT} = Nominal (ON)	1	1.0	1.3	1.6	1.0	1.3	1.6	V
		2,3	1.0	1.3	1.6	-	-	-	V
Shutdown Hysteresis	Difference between voltage threshold of V_{SDI} (ON) and V_{SDI} (OFF)	1	-	0.02	0.2	-	0.02	0.2	V
		2,3	-	0.03	0.2	-	-	-	V
Ripple Rejection ②	f = 1KHz to 10KHz 10mA ≤ I_{OUT} ≤ 1.0A 1.0V = V_{IN}-V_{OUT}	4	20	-	-	20	-	-	dB
		5,6	20	-	-	-	-	-	dB
Phase Margin ②	I_{OUT} = 450mA	4,5,6	30	80	-	30	80	-	degrees
Gain Margin ②	I_{OUT} = 450mA	4,5,6	10	30	-	10	30	-	dB
Equivalent Noise Voltage ②	Referred to Feedback Pin	4,5,6	-	-	50	-	-	50	µVRMS
Thermal Resistance ②	Junction to Case @ 125°C Output Device	-	-	7.3	8.4	-	7.3	9.0	°C/W

NOTES:

① Unless otherwise specified, V_{BIAS} = VIN = 5.0V, R1 = 1.62K, V_{SHUTDOWN} = 0V and I_{OUT} = 10mA. I_{OUT} is subtracted from IQ measurement. See typical application circuit.

② Guaranteed by design but not tested. Typical parameters are representative of actual device performance but are for reference only.

③ Industrial grade devices shall be tested to subgroups 1 and 4 unless otherwise requested.

④ Military grade devices ("H" suffix) shall be 100% tested to subgroups 1, 2, 3 and 4.

⑤ Subgroup 5 and 6 testing available upon request.

⑥ Subgroup 1, 4   TC = +25°C
   Subgroup 2, 5   TC = +125°C
   Subgroup 3, 6   TA = -55°C

⑦ Output current limit is tested with a low duty cycle pulse to minimize junction heating and is dependent on the values of V_{IN}, V_{OUT} and case temperature. See Typical Performance Curves.

⑧ Continuous operation at or above absolute maximum ratings may adversely effect the device performance and/or life cycle.

⑨ Pre and post irradiation limits at 25°C, up to 100Krad TID, are identical unless otherwise specified.

⑩ Reference DLA SMD 5962R05220 for electrical specifications for devices purchased as such.

PIN FUNCTIONS

VIN A,B,C,D,E - These pins provide the input power connection to the MSK5910RH. This is the supply that will be regulated to the output. All five pins must be connected for proper operation.

VBIAS - This pin provides power to all internal circuitry including bias, start-up, thermal limit and overcurrent latch. VBIAS voltage range is 2.9V to 7.5V. VBIAS should be kept greater than or equal to VIN.

GND1 - Internally connected to input ground, these pins should be connected externally by the user to the circuit ground and the GND2 pins.

LATCH - The MSK5910RH LATCH pin is used for both current limit and thermal limit. A capacitor between the LATCH pin and ground sets a time out delay in the event of an over current or short circuit condition. The capacitor is charged to approximately 1.6V from a 7.2µA (nominal) current source. Exceeding the thermal limit will charge the latch capacitor from a larger current source for a near instant shutdown. Once the latch capacitor is charged the device latches off until the latch is reset. Momentarily pull the LATCH pin low, toggle the shutdown pin high then low or cycle the power to reset the latch. Toggling the shutdown pin or cycling the bias power both disable the device during the reset operation (see SHUTDOWN pin description). Pulling the LATCH pin low immediately enables the device for as long as the LATCH pin is held low plus the time delay to re-charge the latch capacitor whether or not the fault has been corrected. Disable the latch feature by tying the LATCH pin low. With the LATCH pin held low the thermal limit feature is disabled and the current limit feature will force the output voltage to droop but remain active if excessive current is drawn.

SHUTDOWN - There are two functions to the SHUTDOWN pin. It may be used to disable the output voltage or to reset the LATCH pin. To activate the shutdown/reset functions the user must apply a voltage greater than 1.3V to the SHUTDOWN pin. The voltage applied to the SHUTDOWN pin can be greater than the input voltage. The output voltage will turn on when the SHUTDOWN pin is pulled below the threshold voltage. If the SHUTDOWN pin is not used, it should be connected to ground.

FB - The FB pin is the inverting input of the internal error amplifier. The non-inverting input is connected to an internal 1.265V reference. This error amplifier controls the drive to the output transistor to force the FB pin to 1.265V. An external resistor divider is connected to the output, FB pin and ground to set the output voltage.

GND2 - Internally connected to output ground, these pins should be connected externally by the user to the circuit ground and the GND1 pins.

VOUT A,B,C,D,E - These are the output pins for the device. All five pins must be connected for proper operation.

OUTPUT CAPACITOR SELECTION

Low ESR output capacitors are required to maintain regulation and stability. Four CWR29FB227 (PN TAZH227K010L) tantalum capacitors in parallel with ceramic decoupling capacitors (0.1µF typical) provides sufficient gain and phase margin for most applications. The maximum ESR specification for the CWR29FB227 capacitor is 180mΩ at 100kHz and is sufficient for many applications. It has been found through full WCCA on the MSK5820RH-1.5 that screening for a maximum ESR of 57mΩ ensures EOL stability criteria to be met for many applications with the most stringent requirements. Analysis of the final design is recommended to ensure stability requirements are met.

POWER SUPPLY BYPASSING

To maximize transient response and minimize power supply transients it is recommended that a 33µF minimum tantalum capacitor is connected between VIN and ground. A 0.1µF ceramic capacitor should also be used for high frequency bypassing.

START UP OPTIONS

The MSK5910RH starts up and begins regulating immediately when VBIAS and VIN are applied simultaneously. Applying VBIAS before VIN starts the MSK5910RH up in a disabled or latched state. When starting in a latched state the device output can be enabled either by pulling the latch pin low to drain the latch capacitor or pulsing the shutdown pin high. The shutdown pulse duration is partially dependent upon the size of the latch capacitor and should be characterized for each application; 30uS is typically adequate for a 1uF latch capacitor at 25°C. A momentary high pulse on the shutdown pin can be achieved using the RC circuit below if VIN rises rapidly. The resistor and capacitor must be selected based on the required pulse duration, the rise characteristic of VIN and the shutdown pin threshold (see shutdown pin threshold and current curves).
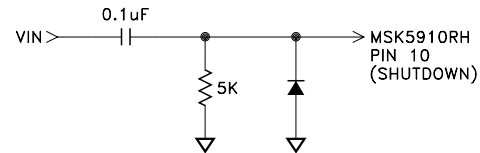
The shutdown pin can be held high and pulled low after VIN comes up or the latch pin held low and released after VIN comes up to ensure automatic startup when applying VBIAS before VIN. Either of the basic circuits below can be adapted to a variety of applications for automatic start up when VBIAS rises before VIN.

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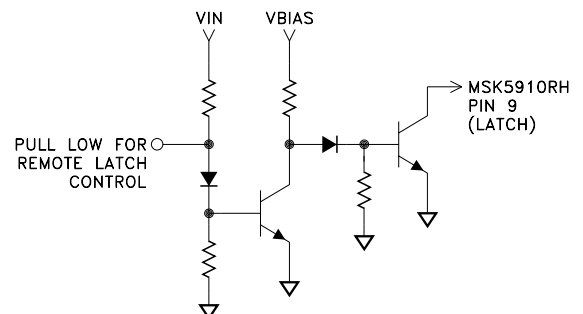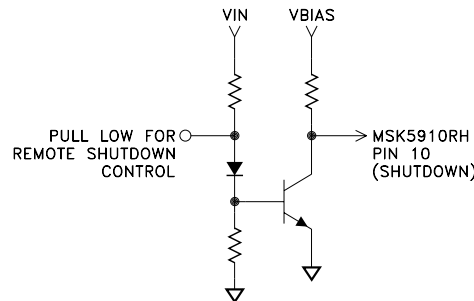
OVERCURRENT LATCH-OFF/LATCH PIN CAPACITOR SELECTION

As previously mentioned, the LATCH pin provides over current/output short circuit protection with a timed latch-off circuit. Reference the LATCH pin description note. The latch off time out is determined with an external capacitor connected from the LATCH pin to ground. The time-out period is equal to the time it takes to charge this external capacitor from 0V to 1.6V. The latch charging current is provided by an internal current source. This current is a function of bias voltage and temperature (see latch charging current curve). For instance, at 25°C, the latch charging current is 7.2µA at VBIAS = 3V and 8µA at VBIAS = 7V.

In the latch-off mode, some additional current will be drawn from the bias supply. This additional latching current is also a function of bias voltage and temperature (see typical performance curves).

The MSK5910RH current limit function is directly affected by the input and output voltages. Custom current limit is available; contact the factory for more information..

THERMAL LIMITING

The MSK5910RH control circuitry has a thermal shutdown temperature of approximately 150°C. This thermal shutdown can be used as a protection feature, but for continuous operation, the junction temperature of the pass transistor must be maintained below 150°C. Proper heat sink selection is essential to maintain these conditions. Exceeding the thermal limit activates the latch feature of the MSK5910RH. See LATCH pin description for instructions to reset the latch or disable the latch feature.

HEAT SINK SELECTION

To select a heat sink for the MSK5910RH, the following formula for convective heat flow may be used.

Governing Equation:

$$T_J = P_D \times (R_{\theta JC} + R_{\theta CS} + R_{\theta SA}) + T_A$$

Where

T_J	=	Junction Temperature
P_D	=	Total Power Dissipation
$R_{\theta JC}$	=	Junction to Case Thermal Resistance
$R_{\theta CS}$	=	Case to Heat Sink Thermal Resistance
$R_{\theta SA}$	=	Heat Sink to Ambient Thermal Resistance
T_A	=	Ambient Temperature

Power Dissipation = (VIN-VOUT) x IOUT

Next, the user must select a maximum junction temperature. The absolute maximum allowable junction temperature is 150°C. The equation may now be rearranged to solve for the required heat sink to ambient thermal resistance ($R_{\theta SA}$).

Example:

An MSK5910RH is connected for V_{IN} = +5V and V_{OUT} = +3.3V. I_{OUT} is a continuous 1A DC level. The ambient temperature is +25°C. The maximum desired junction temperature is +125°C.

$R_{\theta JC}$ = 8.5°C/W and $R_{\theta CS}$ = 0.15°C/W for most thermal greases

Power Dissipation = (5V-3.3V) x (1A)
= 1.7Watts

Solve for $R_{\theta SA}$:

$$R_{\theta SA} = \left[\frac{125°C - 25°C}{1.7W} \right] -8.4°C/W - 0.15°C/W$$

$$= 50.3°C/W$$

In this example, a heat sink with a thermal resistance of no more than 50°C/W must be used to maintain a junction temperature of no more than 125°C.

TYPICAL APPLICATIONS CIRCUIT
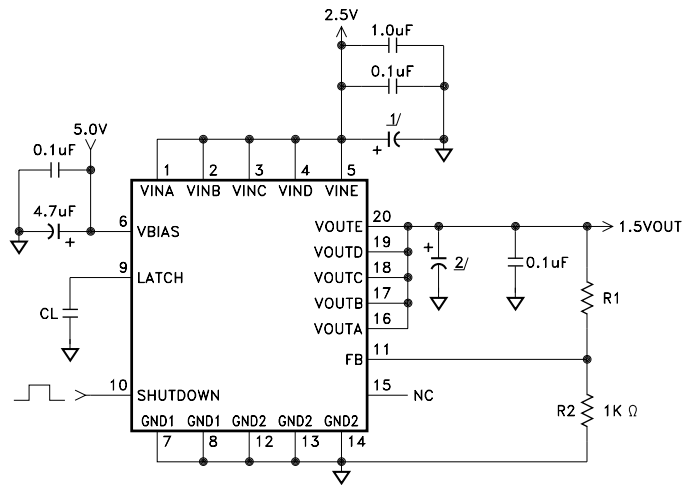
1/ 2 PIECES PART NUMBER
TAZH476K020C (CWR19JC476K)

2/ 4 PIECES PART NUMBER
TAZH227K010L (CWR29FB227K)

$$V_{OUT} = 1.265 (1+R1/R2)$$

OUTPUT VOLTAGE SELECTION

As noted in the above typical applications circuit, the formula for output voltage selection is

$$V_{OUT} = 1.265 \left[1 + \frac{R1}{R2} \right]$$

A good starting point for this output voltage selection is to set R2 = 1K. By rearranging the formula it is simple to calculate the final R1 value.

$$R1 = R2 \left[\frac{V_{OUT}}{1.265} - 1 \right]$$

## APPLICATION NOTES CONT'D

START UP CURRENT

The MSK5910RH sinks increased current during startup to bring up the output voltage. Reference the "Saturated Drive Current vs. Input Voltage" graph in the typical performance curves of this data sheet and the "Understanding Startup Surge Current With RH1573 Based Rad Hard LDO Regulators" application note in the application notes section of the TTM Technologies website for more information.

TOTAL DOSE RADIATION TEST PERFORMANCE

Radiation performance curves for TID testing have been generated for all radiation testing performed by TTM Technologies. These curves show performance trends throughout the TID test process and can be located in the MSK5910RH radiation test report. The complete radiation test report is available in the RAD HARD PRODUCTS section on the TTM Technologies website.

TYPICAL PERFORMANCE CURVES

LATCH PIN LATCH-OFF THRESHOLD vs. VBIAS VOLTAGE
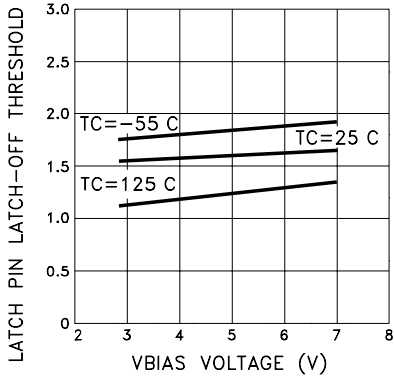
SOA OUTPUT CURRENT vs. VOLTAGE DROP
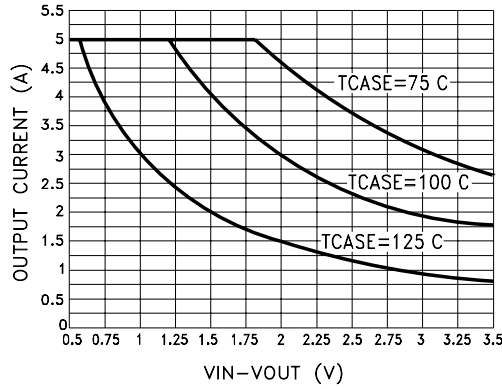
CURRENT LIMIT vs. INPUT VOLTAGE
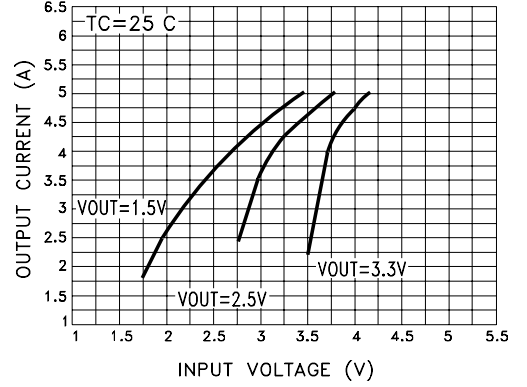
LATCHING CURRENT vs. VBIAS VOLTAGE
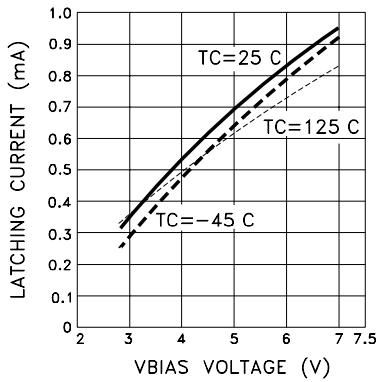
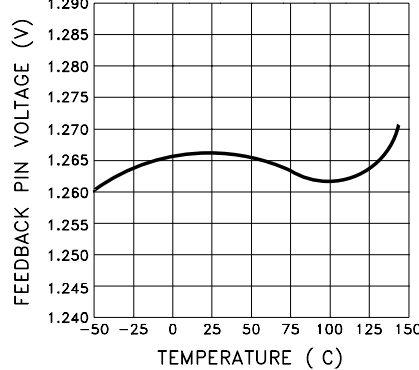### LATCH CHARGING CURRENT vs. VBIAS VOLTAGE
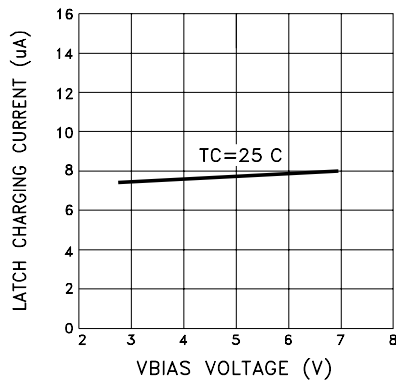
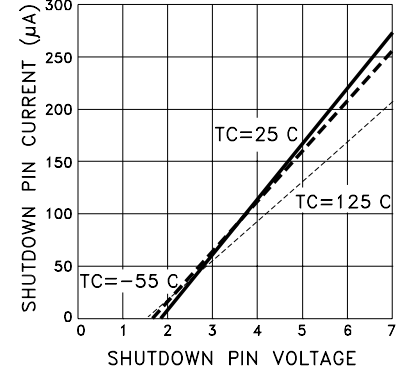### SHUTDOWN VOLTAGE THRESHOLD vs. TEMPERATURE
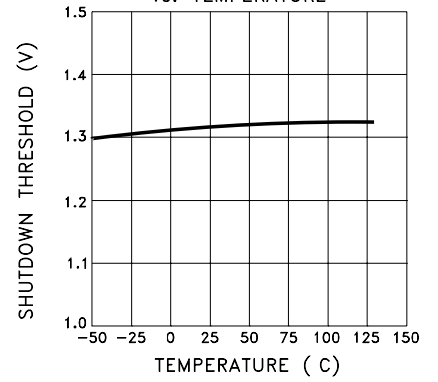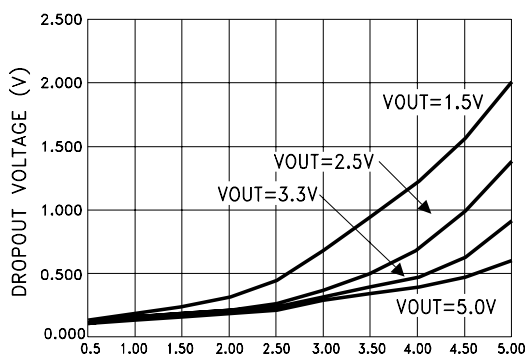
FEEDBACK PIN BIAS CURRENT vs. TEMPERATURE
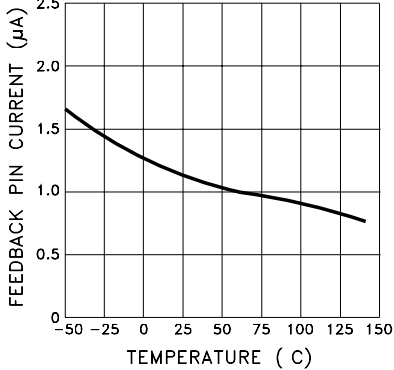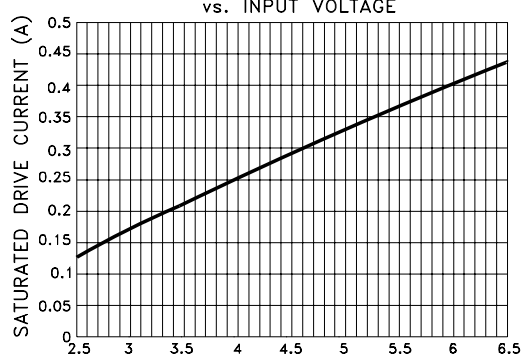
FEEDBACK PIN VOLTAGE vs. TEMPERATURE

SHUTDOWN PIN CURRENT vs. SHUTDOWN PIN VOLTAGE
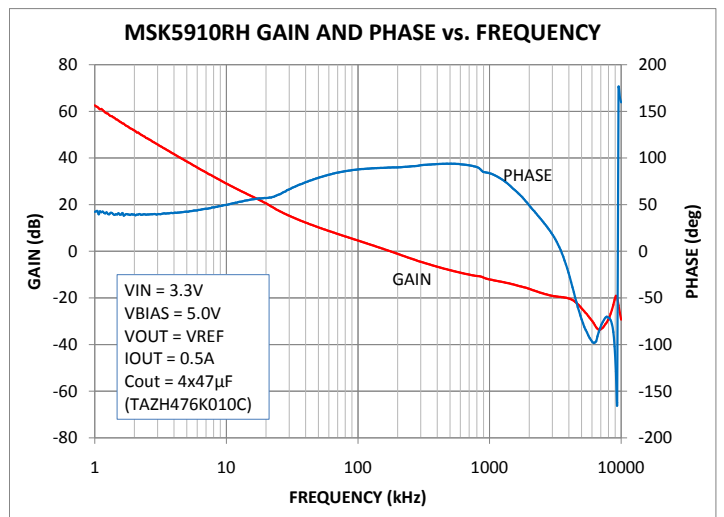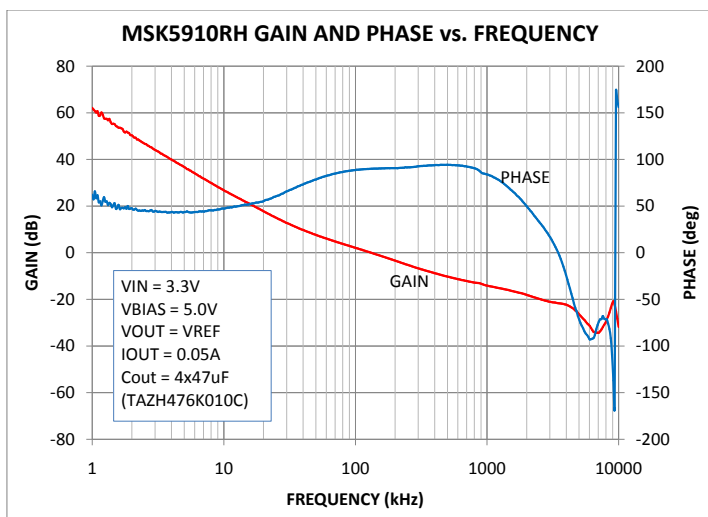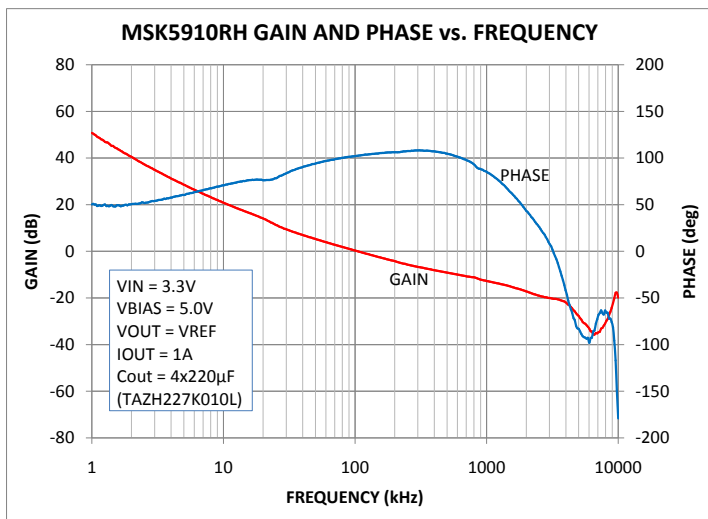
DROPOUT VOLTAGE vs. OUTPUT CURRENT

SATURATED DRIVE CURRENT vs. INPUT VOLTAGE

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GAIN AND PHASE RESPONSE

The gain and phase response curves are for the typical application circuit and are representative of typical device performance, but are for reference only. The performance should be analyzed for each application to insure individual program requirements are met. External factors such as temperature, input and output voltages, capacitors, etc. all can be major contributors. Please consult factory for additional details.

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 0.05A
Cout = 4x220µF
(TAZH227K010L)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 0.5A
Cout = 4x220µF
(TAZH227K010L)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 1A
Cout = 4x220µF
(TAZH227K010L)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 2A
Cout = 4x220µF
(TAZH227K010L)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 0.05A
Cout = 4x47uF
(TAZH476K010C)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 0.5A
Cout = 4x47uF
(TAZH476K010C)

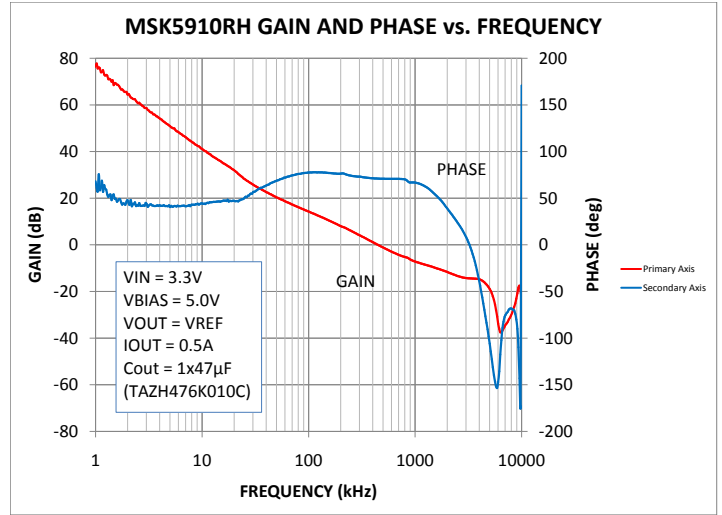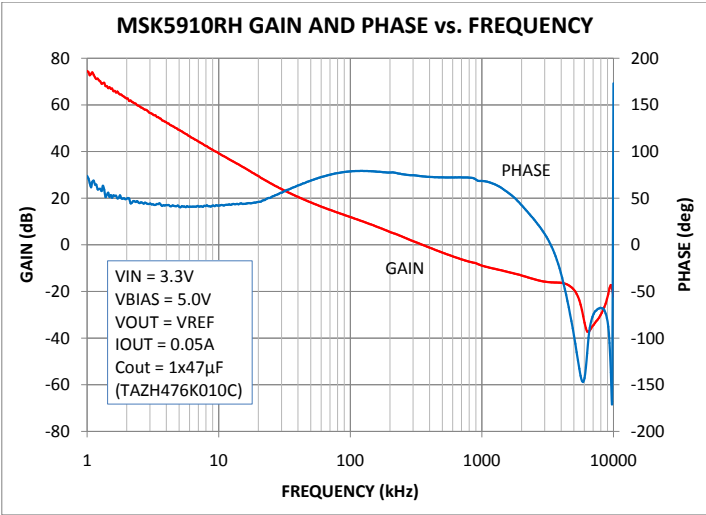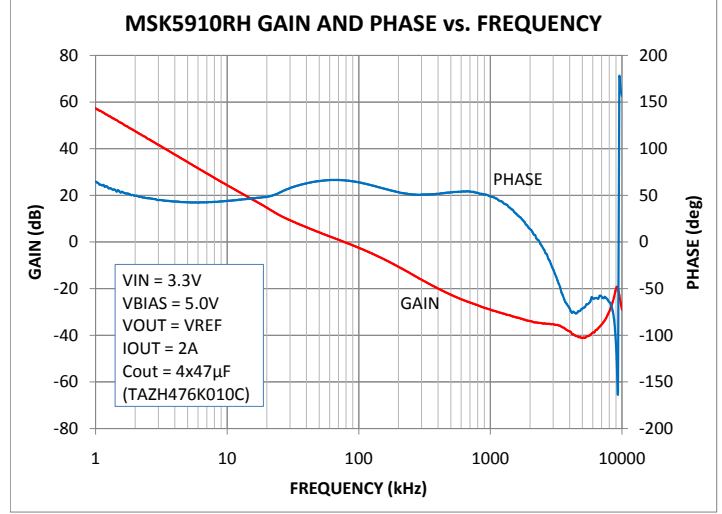8548-11 Rev. T 3/22
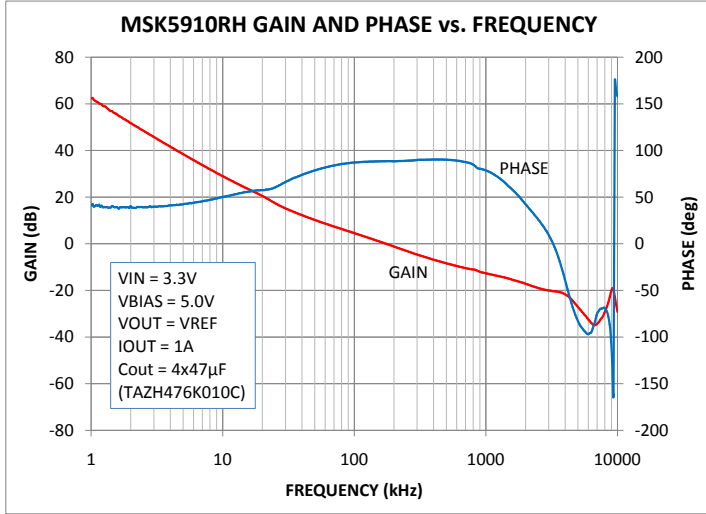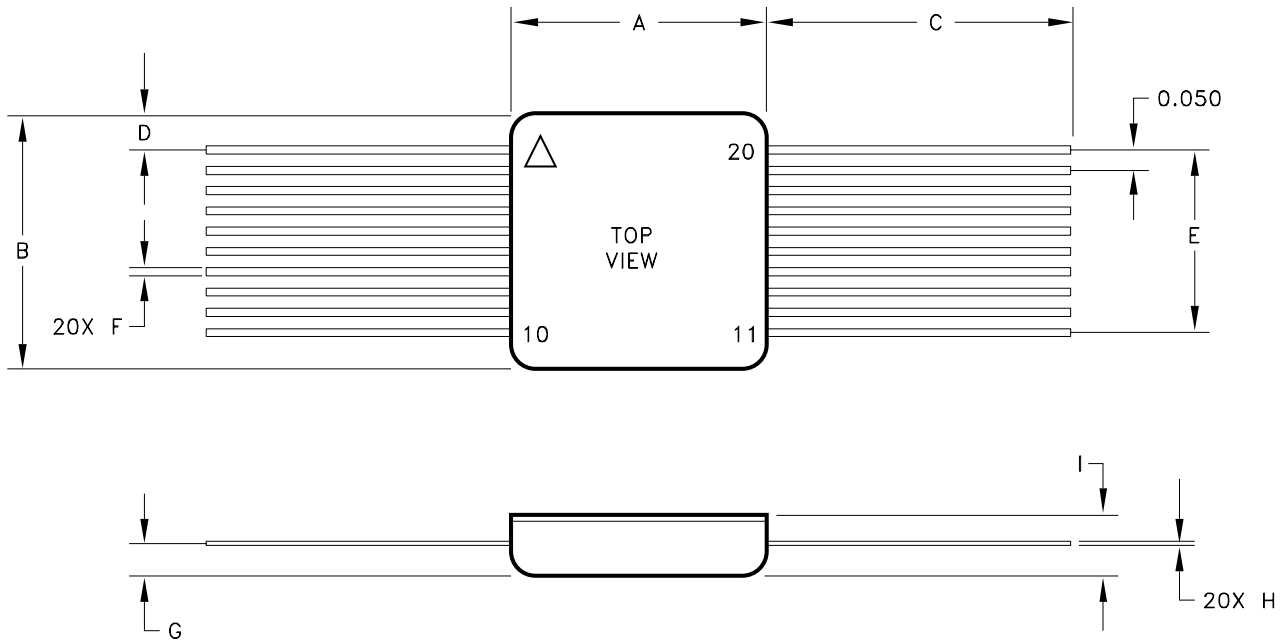
GAIN AND PHASE RESPONSE

The gain and phase response curves are for the typical application circuit and are representative of typical device performance, but are for reference only. The performance should be analyzed for each application to insure individual program requirements are met. External factors such as temperature, input and output voltages, capacitors, etc. all can be major contributors. Please consult factory for additional details.

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 1A
Cout = 4x47µF
(TAZH476K010C)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 2A
Cout = 4x47µF
(TAZH476K010C)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 0.05A
Cout = 1x47µF
(TAZH476K010C)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 0.5A
Cout = 1x47µF
(TAZH476K010C)

Primary Axis
Secondary Axis

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 1A
Cout = 1x47µF
(TAZH476K010C)

MSK5910RH GAIN AND PHASE vs. FREQUENCY

VIN = 3.3V
VBIAS = 5.0V
VOUT = VREF
IOUT = 2A
Cout = 1x47µF
(TAZH476K010C)

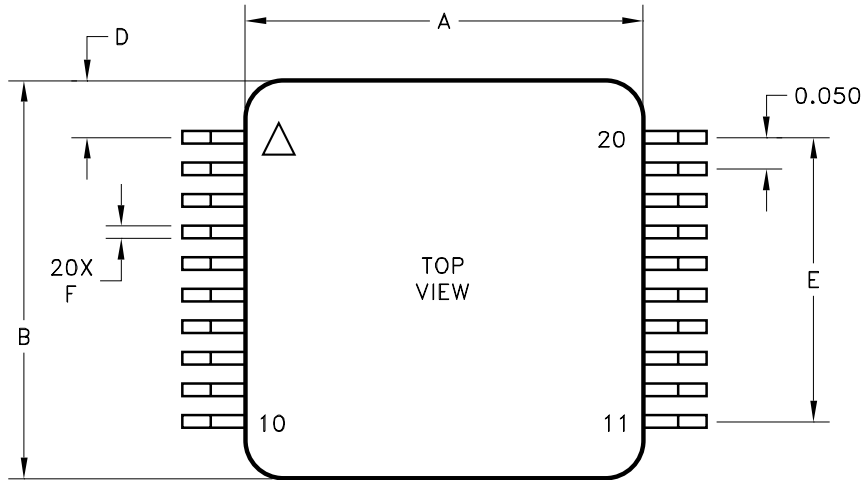| REF	MIN	MAX
A	0.620	0.630
B	0.620	0.630
C	0.500	—
D	0.087 REF.	
E	0.445	0.455
F	0.013	0.017
G	0.076	0.086
H	0.008	0.012
I	0.145	0.155

ESD TRIANGLE INDICATES PIN 1
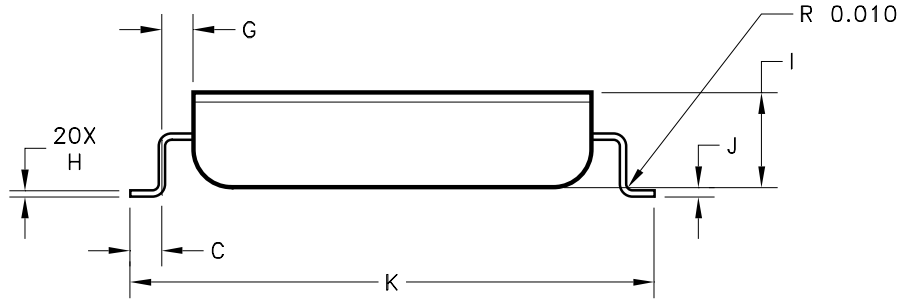WEIGHT = 3.5 GRAMS TYPICAL

ALL DIMENSIONS ARE SPECIFIED IN INCHES

ORDERING INFORMATION

Part Number	Screening Level	LEADS
MSK5910RH	INDUSTRIAL	
MSK5910HRH	MIL-PRF-38534 CLASS H	STRAIGHT
MSK5910KRH	MIL-PRF-38534 CLASS K	
5962R05220	DLA SMD	

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REF	MIN	MAX
A	0.620	0.630
B	0.620	0.630
C	0.045	0.055
D	0.087	REF.
E	0.445	0.455
F	0.013	0.018
G	0.045	0.055
H	0.008	0.012
I	——	0.155
J	0.008	0.018
K	0.805	0.845

TOP VIEW

20X F

20X H

R 0.010

0.050

ESD TRIANGLE INDICATES PIN 1
WEIGHT = 3.3 GRAMS TYPICAL

ALL DIMENSIONS ARE SPECIFIED IN INCHES

ORDERING INFORMATION

Part Number	Screening Level	LEADS
MSK5910RHG	INDUSTRIAL	GULL WING
MSK5910HRHG	MIL-PRF-38534 CLASS H	
MSK5910KRHG	MIL-PRF-38534 CLASS K	
5962R05220	DLA SMD	

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REVISION HISTORY

REV	STATUS	DATE	DESCRIPTION
S	Released	06/14	Add maximum rating for shutdown input and clarify mechanical outline
T	Released	03/22	Remove MIL-PRF-38535, update company name and website

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